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Design of Robust Radar Detectors Through Random Perturbation of the Target Signature

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Abstract—The paper addresses the problem of designing radar detectors more robust than Kelly’s detector to possible mismatches of the assumed target signature, but with no performance degradation under matched conditions. The idea is to model the received signal under the signal-plus-noise hypothesis by adding a random component, parameterized via a design covariance matrix, that makes the hypothesis more plausible in presence of mismatches. Moreover, an unknown power of such component, to be estimated from the observables, can lead to no performance loss, under matched conditions. Derivation of the (one-step) GLRT is provided for two choices of the design matrix, obtaining detectors with different complexity and behavior. A third parametric detector is also obtained by an ad-hoc generalization of one of such GLRTs. The analysis shows that the proposed approach can cover a range of different robustness levels that is not achievable by state-of-the-art with the same performance of Kelly’s detector under matched conditions.

Index Terms—Radar detection, adaptive signal detection, direction-of-arrival estimation, matched filter.

I. INTRODUCTION

A. Notation

Vectors and matrices are denoted by boldface lower-case and upper-case letters, respectively. The symbols $| \cdot |$, $\| \cdot \|$, $\det(\cdot)$, $\text{tr}(\cdot)$, $T$, $^\dagger$, denote modulus value, Euclidean norm, determinant, trace, transpose, and conjugate transpose (Hermitian), respectively. $E[\cdot]$ is the statistical expectation operator. $\mathbb{C}$ is the set of complex numbers and $\mathbb{C}^{N \times M}$ is the Euclidean space of $(N \times M)$-dimensional complex matrices. $0$ is the null vector of proper dimension and $N$ stands for the $N \times N$ identity matrix. $P_u$ is the (orthogonal) projection matrix onto the orthogonal complement of the subspace spanned by the vector $u$. Finally, we write $x \sim \mathcal{CN}_N(0, M)$ if $x$ is an $N$-dimensional complex normal vector with zero mean and (Hermitian) positive definite covariance matrix $M$.

B. Existing Results and Motivation

The well-known problem of detecting the possible presence of a coherent return from a given cell under test (CUT) in range, doppler, and azimuth, is classically formulated as the following hypothesis testing problem:

$$
\begin{align*}
H_0 & : \quad z = n \\
H_1 & : \quad z = \alpha v + n
\end{align*}
$$

where $z \sim \mathbb{C}^{N \times 1}$, $n \sim \mathbb{C}^{N \times 1}$, and $v \sim \mathbb{C}^{N \times 1}$ are the received vector, corresponding noise term, and known space-time steering vector of the useful target echo. In general $N$ is the number of processed samples from the CUT; it might be the number of antenna array elements times the number of pulses [1], [2]. The noise term is often modeled according to the complex normal distribution with zero mean and unknown (Hermitian) positive definite covariance matrix $C$, i.e., $n \sim \mathcal{CN}_N(0, C)$. Modeling $\alpha \in \mathbb{C}$ as an unknown deterministic parameter returns a complex normal distribution for $z$ under both hypotheses; the non-zero mean under $H_1$ makes it possible to discriminate the two hypotheses

$$
\begin{align*}
H_0 & : \quad z \sim \mathcal{CN}_N(0, C) \\
H_1 & : \quad z \sim \mathcal{CN}_N(\alpha v, C)
\end{align*}
$$

In the pioneering paper by Kelly [3], the generalized likelihood ratio test (GLRT) is derived for (2), assuming a set of $K \geq N$ independent and identically distributed training (or secondary) data $r_1, \ldots, r_K$, independent also of $z$, free of target echoes, and sharing with the CUT the statistical characteristics of the noise, is available. In [4] the performance of such a detector is assessed when the actual steering vector is not aligned with the nominal one. The analysis shows that it is a selective receiver, i.e., it may have excellent rejection capabilities of signals arriving from directions different from the nominal one. A selective detector is desirable for target localization. Instead, when a radar is working in searching mode, a certain level of robustness to mismatches is preferable. For this reason, many works have addressed the problem of enhancing either the selectivity or the robustness of radar detectors to mismatches. In particular, the adaptive matched filter (AMF) [5], which solves (2) following a two-step approach, is a prominent example of robust detector, while the adaptive coherence estimator (ACE, also known as adaptive normalized matched filter) [6] is another example of selective receiver. Other relevant examples of selective receivers are obtained by solving a modified hypothesis testing problem that assumes the presence of a (fictitious) coherent signal under the noise-only ($H_0$) hypothesis to make it more plausible in presence of signal mismatches [7]–[9]. A family of receivers, obtained by inserting a nonnegative parameter in the original Kelly’s detector, has been proposed by Kalson in [10]; such a parameter, indicated as $\beta$ in the following, allows one to control the degree to which mismatched signals are rejected, so
obtaining behaviors in between the AMF and Kelly’s detector. A different tunable receiver called KWA has been proposed in [11], which encompasses as special cases Kelly’s and W-ABORT detectors [9]. Although designed for enhanced selectivity, for values of its tunable parameter $\gamma$ smaller than $1/2$ it behaves as a robust detector, reaching the energy detector for $\gamma = 0$. Tunable receivers have also been proposed in [12], [13]. Other approaches, as for instance those based on the cone idea, can guarantee an increased robustness at the price of a certain loss under matched conditions [14], [15]. A robust two-stage detector obtained by cascading a GLRT-based subspace detector and the Rao test has been proposed in [16]. Robust and selective detectors have also been considered in the context of subspace detection [17]–[19]. Multidimensional/multichannel signal detection in homogeneous or partially homogeneous Gaussian disturbance (with unknown covariance matrix and unknown structured deterministic interference) is considered in [20], [21].

The on the other hand, modeling $\alpha \in \mathbb{C}$ as a complex normal random variable with zero mean and variance $E[|\alpha|^2] = E[|\alpha|^2]$, returns a zero-mean complex normal distribution for $z$ under both hypotheses; hence

$$\begin{align*}
H_0 & : z \sim \mathcal{CN}_N(0, C) \\
H_1 & : z \sim \mathcal{CN}_N(0, C + \alpha^2 vv^T).
\end{align*}$$

This is a “second-order” approach to target modeling, i.e., the presence of a useful signal ($H_1$ hypothesis) is modeled in terms of a modification of the noise covariance matrix, instead of appearing in the mean as conversely for the more classical problem (2). Interesting properties in terms of either rejection capabilities or robustness to mismatches on the nominal steering vector can be obtained by considering a random (instead of deterministic) target signal, depending on the way problem (3) is solved and possibly on the presence of a fictitious signal under $H_0$ [22], [23].

In this paper, we investigate the potential of a detector that solves the following new hypothesis testing problem

$$\begin{align*}
H_0 & : z = n \\
H_1 & : z = \alpha v + \theta + n
\end{align*}$$

where $\alpha$ is an unknown deterministic parameter and $\theta \sim \mathcal{CN}_N(0, \nu \Sigma)$ represents a random component. The goal is to obtain a detector that exhibits the same probability of detection ($P_d$) of Kelly’s GLRT under matched conditions, but is more robust than the latter to mismatches between the nominal steering vector and the actual one. In the existing literature, the “win-win” situation in which robustness is achieved without any loss under matched conditions has been obtained so far through careful parameter setting of tunable receivers, notably the mentioned Kalson’s and KWA detectors; here we propose a different approach which, remarkably, can cover levels of robustness that are not possible with such state-of-the-art receivers. The idea in (4) is in fact to add to the $H_1$ hypothesis a signal $\theta$ that makes $H_1$ more plausible, hence hopefully the detector more robust to mismatches on the nominal steering vector $v$. To this aim, a design matrix $\Sigma$ is considered, multiplied by an unknown factor $\nu$; in doing so, in case of matched signature no component will be likely found along $\Sigma$ and the conventional case is recovered, i.e., the estimated value for $\nu$ will be likely zero; conversely, if the mismatch causes some leakage of the signal that is captured by $\Sigma$, the detector would tend to decide for $H_1$ more likely.

If we suppose that $\theta$ is independent of $n$, the resulting hypothesis testing problem turns out to be

$$\begin{align*}
H_0 & : z \sim \mathcal{CN}_N(0, C) \\
H_1 & : z \sim \mathcal{CN}_N(\alpha v, C + \nu \Sigma).
\end{align*}$$

In summary, the contribution of this paper is threefold:

- A new hypothesis test (5) for radar detection is introduced, which is a possible generalization of (2) and (3).
- The (one-step) GLRT is derived for two choices of the design matrix $\Sigma$. A third parametric detector is obtained by an ad-hoc generalization of one of such GLRTs. Two-step GLRT-based detectors have been presented in our preliminary work [24]. The detectors derived here have different complexity and behavior.

- A thorough analysis of the proposed detectors is performed, also deriving closed-form expressions for the $P_{fa}$ for two of the detectors and showing that they have the desirable constant false alarm rate (CFAR) property. The performance assessment reveals that the proposed approach can guarantee negligible loss under matched conditions with respect to Kelly’s detector while providing diversified degrees of robustness depending on chosen parameters.

The paper is organized as follows: Section II is devoted to the derivation of the GLRTs (with some proofs and lengthy manipulations in the Appendices). Section III addresses the analysis of the detectors. We conclude in Section IV.

II. GLRTs FOR POINT-LIKE TARGETS

In this section, we derive robust detectors employing the GLRT. To this end, we consider the following binary hypothesis testing problem

$$\begin{align*}
H_0 & : z \sim \mathcal{CN}_N(0, C) \\
H_k & : z \sim \mathcal{CN}_N(0, C + \nu \Sigma) \\
\end{align*}$$

where we recall that the training data $\mathbf{r}_k$ form a set of $K \geq N$ independent (and identically distributed) vectors, independent also of $z$, free of target echoes, and sharing with the CUT the statistical characteristics of the noise. The positive definite matrix $\Sigma$, $\nu \geq 0$, and $\alpha \in \mathbb{C}$ are unknown quantities while $v \in \mathbb{C}^{N \times 1}$ is a known vector. As to the (Hermitian) positive semidefinite matrix $\Sigma$, it might be either known or unknown. In the former case, $\Sigma$ reflects our knowledge about the random variations of the target steering vector around its nominal value $v$. On the contrary, an unknown $\Sigma$ implies that no specific knowledge is available. In such a case, obviously $\Sigma$ cannot be estimated from a single snapshot (data from the CUT) and therefore one needs to make some further assumptions. The latter are mostly dictated by pragmatism so that the detection problem remains identifiable and mathematically tractable. Indeed, $\Sigma$ should be viewed as a means to model the structure of the variations around $v$ while $\nu$ captures their amplitudes. In any case, one has to look

\footnote{The condition $K \geq N$ ensures that the sample covariance matrix based on training data has full rank with probability one [3].}
at the ultimate performance for judging the goodness of each choice. In the sequel, we will assume that either \( \Sigma \) is a known rank-one matrix or \( \Sigma = C \) (so that, although unknown, it does not introduce new unknowns).

A. Case 1: Derivation of the GLRT for \( \Sigma = uu^\dagger \)

In this section, we address the case that \( \Sigma \) is a rank-one matrix. First observe that the case \( \Sigma = vv^\dagger \), which enforces the target signature in both mean and covariance, returns Kelly’s detector, as discussed later in this section. We are instead more interested in a vector \( u \) aimed at making the \( H_1 \) hypothesis more plausible under mismatches. Intuitively, a design matrix \( \Sigma = v_m v_m^\dagger \), where \( v_m \), is a slightly mismatched steering vector, is a possible way to introduce some robustness without appreciable loss under matched conditions; nonetheless, other choices are possible, hence we provide a derivation for generic \( \Sigma = uu^\dagger \).

The corresponding GLRT is given by

\[
\Lambda(z, S) = \max_{C > 0} \max_{\alpha > 0} \max_{u \in \mathbb{C}^N} f_1(z, S | C, \nu, \alpha) \frac{H_1}{H_0} \cdot \eta
\]

where

\[
f_1(z, S | C, \nu, \alpha) = \frac{e^{-u \left[ (\nu u u^\dagger + C)^{-1} (z - \alpha v) (z - \alpha v)^\dagger + C^{-1} S \right]}}{\pi^{N(K+1)} \det(C) \det(\nu u u^\dagger + C)}
\]

and

\[
f_0(z, S | C) = \frac{1}{\pi^{N(K+1)} \det(C)} e^{-u \left[ C^{-1} (z z^\dagger + S) \right]}
\]

are the probability density functions (PDFs) of \( z, r_1, \ldots, r_K \) under \( H_1 \) and \( H_0 \), respectively, and \( S/K \) the sample covariance matrix computed on the secondary data, i.e.,

\[
S = \sum_{k=1}^K r_k r_k^\dagger.
\]

As to \( \eta \), it is the detection threshold to be set according to the desired probability of false alarm (\( P_{fa} \)). Maximization over \( C \) of the likelihood under \( H_0 \) can be performed as in [3]; in fact, the maximizer of the likelihood is given by

\[
\hat{C}_0 = \frac{1}{K + 1} [zz^\dagger + S];
\]

It follows that

\[
f_0(z, S | \hat{C}_0) = \left( \frac{K + 1}{e \pi} \right)^{N(K+1)} \left[ \det(z z^\dagger + S) \right]^{-(K+1)}.
\]

Computation of the numerator of the GLRT and, in particular, maximization of \( f_1(z, S | C, \nu, \alpha) \) with respect to \( C \) is less standard. However, it can be conducted exploiting the following result (a special case of the one in [23]).

Proposition 1: Assume that \( u \) is a unit-norm vector (and that \( K \geq N \)). Then, for the likelihood under the \( H_1 \) hypothesis, given by equation (7), the following equality holds true

\[
l_{\max} = \max_{C > 0, \alpha > 0, \nu \in \mathbb{C}} f_1(z, S | C, \nu, \alpha)
\]

\[
= \max_{b > 0} \max_{\alpha \in \mathbb{C}} \left[ \frac{(K + 1)}{e \pi} \right]^{N(K+1)} \left( 1 + b \right)^{-1} \frac{\left[ \det(S) \left( 1 + (z - \alpha v)^\dagger S^{-1} (z - \alpha v) \right) \right]^{K+1}}{u^\dagger \left( S + (1 + b)^{-1} (z - \alpha v) (z - \alpha v)^\dagger \right)^{-1} u}
\]

\[
\times \left[ \frac{\left( S + (z - \alpha v) (z - \alpha v)^\dagger \right)^{-1}}{u^\dagger \left( S + (z - \alpha v) (z - \alpha v)^\dagger \right)^{-1} u} \right]^{K+1}
\]

where \( b = \nu u u^\dagger - u \).

Proof: The result is a special case of the one in [23] letting therein, in equation (39), \( R = C, T_p = 1, T_a = K \) (hence, \( T_a = K + 1 \), \( M = N \), \( X = z - \alpha v \), \( S_y = S, v = u, P = \nu \).

In the following, we will denote by \( l(b, \alpha) \) the partially-compressed likelihood under \( H_1 \), i.e.,

\[
l(b, \alpha) = \frac{(K + 1)}{e \pi} \left( 1 + b \right)^{-1} \frac{\left[ \det(S) \left( 1 + (z - \alpha v)^\dagger S^{-1} (z - \alpha v) \right) \right]^{K+1}}{u^\dagger \left( S + (1 + b)^{-1} (z - \alpha v) (z - \alpha v)^\dagger \right)^{-1} u}
\]

As shown in Appendix A, it can also be re-written as

\[
l(b, \alpha) = \frac{(K + 1)}{e \pi} \left( 1 + b \right)^{-1} \frac{\left[ \det(S) \left( 1 + b + \| z - \alpha \hat{v} \|^2 \right) \right]^{K+1}}{u^\dagger \left( S + \| P_u^\dagger (z - \alpha \hat{v}) \|^2 \right)^{K+1}}
\]

where \( z = S^{-1/2} z \), \( \hat{v} = S^{-1/2} v \), and \( \hat{u} = S^{-1/2} u \) are the “whitened” versions of \( z, v \), and \( u \), respectively, and we recall that \( P_u^\dagger \) is the (orthogonal) projection matrix onto the orthogonal complement of the subspace spanned by \( \hat{u} \).

For \( u = v \), \( l(b, \alpha) \) can be easily maximized and the GLRT is Kelly’s detector (ref. Appendix A). If, instead, \( u \neq v \), maximization of the above partially-compressed likelihood with respect to \( \alpha \in \mathbb{C} \) can be restricted to a proper “segment” of the complex plane. In fact, the following result holds true.

Proposition 2: Let \( P_u^\dagger \hat{v} \neq 0 \). The maximum of the function \( l(b, \alpha) \) with respect to \( \alpha \in \mathbb{C} \), given \( b \geq 0 \), is attained over the segment whose endpoints are \( \alpha_1 \) and \( \alpha_2 \), defined as

\[
\alpha_1 = \arg \min_{\alpha} \| \hat{z} - \alpha \hat{v} \|^2 \quad \alpha_2 = \arg \min_{\alpha} \| P_u^\dagger (\hat{z} - \alpha \hat{v}) \|^2
\]

Proof: See Appendix A.

From the proposition above, \( \alpha \) can be re-written as

\[
\alpha(t) = t \alpha_1 + (1 - t) \alpha_2, \quad t \in [0, 1]
\]

Accordingly, we have

\[
|\alpha(t) - \alpha_1|^2 = t|\alpha_1 - (1 - t) \alpha_2 - \alpha_1|^2 = (1 - t)^2 |\alpha_2 - \alpha_1|^2,
\]

\[
|\alpha(t) - \alpha_2|^2 = |t \alpha_1 + (1 - t) \alpha_2 - \alpha_2|^2 = t^2 |\alpha_1 - \alpha_2|^2;
\]

\[
\alpha_1 = \arg \min_{\alpha} \| \hat{z} - \alpha \hat{v} \|^2
\]

This amounts to assuming that \( z = (\alpha + \epsilon) v + \eta \) where \( \epsilon \) is a zero-mean complex normal random variable with unknown power \( \nu \), independent of \( \eta \).
then, using equations (A.16), (A.18), and (A.19), together with the above two equations, yields

\[
l(b, \alpha(t)) = \left( \frac{K + 1}{\pi \epsilon} \right)^{N(K+1)} \frac{1}{\det^{K+1}(S)} l_1 \left( b, \| \tilde{z} - \alpha(t) \tilde{v} \|^2 \right) l_2 \left( b, \| P_u^\perp (\tilde{z} - \alpha(t) \tilde{v}) \|^2 \right)
\]

\[
= \left( \frac{K + 1}{\pi \epsilon} \right)^{N(K+1)} \frac{1}{\det^{K+1}(S)} \times l_1 \left( b, \tilde{z} \right) P_u^\perp P \tilde{z} + (1 - t)^2 |\alpha_2 - \alpha_1|^2 \tilde{v} \|	ilde{v} \|^2
\]

\[
\times l_2 \left( b, \tilde{z} \right) P_u^\perp P \tilde{z} + t^2 |\alpha_2 - \alpha_1|^2 \tilde{v} \|	ilde{v} \|^2 \right)
\]

with \( l_1(\cdot, \cdot) \) and \( l_2(\cdot, \cdot) \) given in equation (A.17). Summarizing, the GLRT can be written as

\[
\begin{align*}
\max_{b \geq 0} \max_{t \in [0,1]} l(b, \alpha(t)) & \quad H_1 > \eta \quad H_0
\end{align*}
\]

and also as

\[
\begin{align*}
l_1 \left( \tilde{z} \right) P_u^\perp P \tilde{z} + (1 - t)^2 R^2 \tilde{v} \|	ilde{v} \|^2
\end{align*}
\]

\[
\times l_2 \left( \tilde{z} \right) P_u^\perp P \tilde{z} + t^2 |\alpha_2 - \alpha_1|^2 \tilde{v} \|	ilde{v} \|^2 \right)
\]

where \( R^2 = |\alpha_1 - \alpha_2|^2 \). Equation (10) is a computationally more convenient alternative for the GLRT in the case \( \Sigma = uu^\dagger \), with \( u \) any chosen vector. This rewriting is in fact less demanding than using \( l(b, \alpha) \) for the numerator (partially-compressed likelihood under \( H_1 \)): in particular, the maximization over \( \alpha \in C \) of \( l(b, \alpha) \) (e.g., using equation (9)) would require a complex-valued (two-dimensional) unbounded optimization instead of the simple scalar search of \( t \) in the finite interval \( [0,1] \) appearing in (10). Moreover, it is reasonable to constraint \( b \) to belong to a finite interval, say \( [0, b_{\text{max}}] \). This further reduces the ultimate complexity and makes it possible to implement the detector using a numerical algorithm with box constraints or, alternatively, a grid search.

\[\text{B. Case 2: Derivation of the GLRT for } \Sigma = C\]

In this case \( \Sigma \) is unknown, but equal to the covariance matrix of the noise \( C \). Clearly, since the latter is unrelated to the sources of error producing the steering vector mismatch that we hope to capture through the random term \( \theta \), a direct physical meaning for \( \Sigma = C \) is missing; however, such a choice is mathematically convenient and leads to a detector with sound properties, as shown later. Also, if \( C \) were known, the GLRT would process the whitened received signal and the choice \( \Sigma = C \) would be tantamount to modeling the random component to be added to the (whitened) steering vector as a white term. This is reasonable when no a-priori knowledge on the nature of the mismatch is available, hence \( \Sigma = C \) is actually a conservative (less informative) choice.

\[\text{3 Please also recall the general rationale of test (4) and the discussion at the beginning of Section II.}\]

The corresponding GLRT is

\[
\Lambda(z, S) = \frac{\max_{C > 0} \max_{\nu > 0} \max_{\alpha \in C} f_1(z, S(C, \nu, \alpha)) \frac{H_1}{H_0}}{f_0(z, S|C)} \leq \eta
\]

(11)

where

\[
f_1(z, S|C, \nu, \alpha) = \frac{e^{-\nu \left[ C^{-1} \left( \frac{1}{\epsilon^2} (z - \alpha v)(z - \alpha v)^\dagger + S \right) \right]}}{\pi^{N(K+1)}(1 + \nu)^N \det^{K+1}(C)}
\]

is the PDF of \( z, r_1, \ldots, r_K \) under \( H_1 \) while the denominator of (11) is given by (8). Again \( \eta \) is the detection threshold for the desired \( P_{fa} \). We also recall that \( S \) is \( K \) times the sample covariance matrix computed on the secondary data.

Maximization over \( C \) of the likelihood under \( H_1 \) can be performed as in [3]; in fact, we have that

\[
\hat{C}_1(\nu, \alpha) = \frac{1}{K + 1} \left[ \frac{1}{1 + \nu} (z - \alpha v)(z - \alpha v)^\dagger + S \right]
\]

Thus, the partially-compressed likelihood under \( H_1 \) becomes

\[
f_1(z, S|\hat{C}_1(\nu, \alpha), \nu, \alpha) = \left( \frac{K + 1}{\epsilon^2} \right)^{N(K+1)} \left[ \frac{1}{1 + \nu} \right]^{N-1} \left[ \det \left( 1 + \frac{1}{1 + \nu} (z - \alpha v)(z - \alpha v)^\dagger + S \right) \right]^{-\frac{1}{K+1}}
\]

Plugging the above expressions into equation (11) yields

\[
\Lambda_{\pi^{\infty}}(z, S) = \frac{1 + \|\tilde{z}\|^2}{\min_{\nu > 0, \alpha \in C \left[ 1 + \nu \right]}^{\infty} \left[ 1 + \frac{1}{1 + \nu} \| \tilde{z} - \alpha \tilde{v} \|^2 \right]}
\]

Moreover, minimization over \( \alpha \) leads to (25)

\[
\Lambda_{\pi^{\infty}}(z, S) = \frac{1 + \|\tilde{z}\|^2}{\min_{\nu > 0 (1 + \nu)}^{\infty} \left[ 1 + \frac{1}{1 + \nu} \| \tilde{z} \|^2 \right]}
\]

Minimization of denominator with respect to \( \nu \) can be easily accomplished using the following proposition.

**Proposition 3:** The function

\[
f(\nu) = (1 + \nu) \pi^{\nu} \left( 1 + \frac{a}{1 + \nu} \right), \quad a > 0, K \geq N
\]

admits a unique minimum over \([0, +\infty]\) at

\[
\hat{\nu} = \begin{cases} (\frac{K + 1}{N} - 1) a - 1, & (\frac{K + 1}{N} - 1) a - 1 > 0 \\ 0, & \text{otherwise} \end{cases}
\]

given by

\[
f_{\min} = \begin{cases} (\frac{K + 1 - N}{N} a)^{\frac{N}{\nu}} (\frac{K + 1}{K + 1 - N} a), & K + 1 - N > 0 \\ 1 + a, & \text{otherwise} \end{cases}
\]

**Proof:** See Appendix B.

Thus, it follows that the GLRT can be written as

\[
\Lambda_0(z, S) = \frac{\Lambda_{\pi^{\infty}}(z, S)}{H_1} \leq \eta \quad H_0
\]

(12)
where

\[
\Lambda_0(z, S) = \begin{cases} 
\left(1 + \|\hat{z}\|^2\right) \left(1 - \frac{1}{\zeta}\right) \frac{1}{1 - \zeta}, & \|P_0\hat{z}\|^2 > \frac{1}{\zeta - 1} \\
\left[(\zeta - 1) \|P_0\hat{z}\|^2\right]^{\frac{1}{\zeta}}, & \|P_0\hat{z}\|^2 < \frac{1}{\zeta - 1} \\
\frac{1 + \|\hat{z}\|^2}{1 + \|P_0\hat{z}\|^2}, & \text{otherwise}
\end{cases}
\]

\[\zeta = \frac{K + 1}{N},\] and \(\eta\) denotes a modification of the original threshold. The \(P_{fa}\) of the test and the characterization of the decision statistic under \(H_1\) can be obtained in closed form, and are reported in Appendix B. Remarkably, in the appendix it is shown that the detector has the desirable CFAR property.

Notice that, conditionally to \(\zeta\) in (13) with

\[\zeta = \frac{K + 1}{N}(1 + \epsilon), \quad \epsilon \geq 0.\] (14)

We made this ad-hoc substitution also on the decision statistic, obtaining the following parametric detector that thus generalizes (12) through (14):

\[\Lambda_\epsilon(z, S) = \frac{H_1}{H_0} \geq \eta\] (15)

where

\[
\Lambda_\epsilon(z, S) = \begin{cases} 
\left(1 + \|\hat{z}\|^2\right) \left(1 - \frac{1}{\zeta_\epsilon}\right) \frac{1}{1 - \zeta_\epsilon}, & \|P_0\hat{z}\|^2 > \frac{1}{\zeta_\epsilon - 1} \\
\left[(\zeta_\epsilon - 1) \|P_0\hat{z}\|^2\right]^{\frac{1}{\zeta_\epsilon}}, & \|P_0\hat{z}\|^2 < \frac{1}{\zeta_\epsilon - 1} \\
\frac{1 + \|\hat{z}\|^2}{1 + \|P_0\hat{z}\|^2}, & \text{otherwise}
\end{cases}
\]

Such parametric detector encompasses detector (12) for \(\epsilon = 0\) and eventually \(\zeta_\epsilon = \zeta\). Moreover, we will show that detector (15) can outperform detector (12) in terms of robustness, without incurring additional loss under matched conditions. Finally, a closed form for its \(P_{fa}\) is given in equation (B.24) of Appendix B, which allows one to set the threshold to guarantee the desired \(P_{fa}\) for any choice of \(\epsilon\); for convenience, the curves for some relevant values of \(\epsilon\) are drawn in Fig. 1, also revealing that the \(P_{fa}\) does not significantly change with \(\epsilon\). In Appendix B it is furthermore proven that the proposed detector (15) (hence also (12)) possesses the CFAR property, i.e., its \(P_{fa}\) does not depend on any unknown parameter (including the unknown noise covariance matrix \(C\)); additionally, it is proven that its performance in terms of \(P_{fa}\) depends only on the SNR (equation (B.25)) and the cosine squared \(\cos^2 \theta\) (equation (B.26)) between the nominal steering vector and the actual one, which are the same performance parameters of Kelly’s, AMF, Kalson’s, and KWA detectors. This remarkable fact corroborates the soundness and convenience of the choice \(\Sigma = C\). It also means that \(P_{fa}\) does not depend directly on noise-related parameters, namely the clutter power and correlation, but only through SNR and \(\cos^2 \theta\); as a consequence, the figures we will provide in Section III (\(P_{fa}\) vs SNR for different \(\cos^2 \theta\)) are quite representative of the general performance, despite they are obtained for a specific choice of \(C\).

C. Discussion

It is important to stress that the derivation of the GLRT for \(\Sigma = C\) and hence, the result is different from the detector proposed in [6]. More precisely, considering the derivation of the ACE as the GLRT to detect a possible coherent signal in presence of partially-homogeneous noise, given in [6], we highlight that therein the parameter \(\nu\) represents the possible mismatch between the power of the noise in the data under test and that of secondary data. Herein, instead, \(\nu\) enters the characterization of a possible random mismatch between the actual and the nominal useful signal. For this reason, we have \(1 + \nu\) in place of \(\nu\) under \(H_1\) (\(\nu C\) stands for the covariance matrix of the random term) and, in addition, \(\nu\) is not present under the \(H_0\) hypothesis. This difference in the formulation of the hypothesis testing problem makes our detector more robust than Kelly’s detector [3] in presence of mismatched signals while it is well-known that the ACE is more selective than the latter (under the same operating conditions).

Table I reports a summary of the proposed detectors for test (5), also in comparison with the classical approach (test (2)).

III. Performance Analysis

We assume a time steering vector, i.e., \(v = [1, e^{i2\pi f_d}, \ldots, e^{i2\pi (N-1) f_d}]^T\), with \(N = 16\) and normalized Doppler frequency \(f_d = 0.08\), a small value such that the target competes with low pass clutter. The target amplitude \(\alpha\) is generated deterministically according to the SNR defined in equation (B.25). To assess the robustness of the proposed detectors, we simulate a target with a mismatched steering vector, say \(p\), having normalized Doppler frequency \(f_d + \delta_f\) with \(\delta_f \in \{0.2/N, 0.4/N\}\).

For completeness, we mention that [24] provides the general derivation (for any \(\Sigma\)) and a closed-form for \(\Sigma = C\) following the two-step GLRT approach, namely deriving the GLRT for known \(C\) and then using as replacement an estimate \(\hat{C}\) (from secondary data); such detectors are however less powerful than one-step GLRTs. In Appendix C we show that for \(\Sigma = \nu v^H\) the two-step GLRT is equivalent to the AMF, whereas the proposed one-step GLRT for the same choice of \(\Sigma\) is equivalent to Kelly’s detector (ref. Section II-A).
In the following we assess the performance of the proposed (one-step) GLRTs, considering both $\Sigma = \mathbf{v}_m^H \mathbf{v}_m$ and $\Sigma = \mathbf{C}$, and the parametric detector (15). The former detector assumes $\mathbf{v}_m$ a slightly mismatched version of $\mathbf{v}$, in particular of $0.03/N$, and for the optimization with respect to $\beta$ we adopt a logarithmic search in the span $[0, b_{\text{max}}]$; as regards the parametric detector (15), we consider $\epsilon = 0.1$ and $\epsilon = 0.2$. It will be apparent from the simulations below that such choices guarantee negligible loss with respect to Kelly’s detector under matched conditions.

A comparison is performed against the natural competitors for the problem at hand, that is the Kalson’s [10] and KWA [11] tunable receivers. Notice that for both such competitors, the level of robustness increases as the value of their tunable parameter decreases. For a fair comparison, only values that ensure practically the same performance of Kelly’s detector under matched conditions are considered, in particular $\beta = 0.5$ and $\beta = 0.8$ for Kalson’s detector and $\gamma = 0.45$ and $\gamma = 0.4$ for the KWA. For reference, the AMF [5] is also reported in all figures, because it is a well-known robust detector; however, it is worth remarking that it is not always a competitor, since it generally experiences a performance loss under matched conditions with respect to Kelly’s detector.

A first set of simulations (Figs. 2–4) refers to $P_{fa} = 10^{-4}$. We assess performance by Monte Carlo simulation with 100/$P_{fa}$ independent trials to set the thresholds and 10$^3$ independent trials to compute the $P_{fs}$ (the probabilities to decide for $H_1$ when a useful signal is present). Results under matched conditions are reported in Fig. 2 for different values of the number of secondary data, namely $K = 24, 32, 40$. Results under mismatched conditions are reported in Fig. 3 for different values of the number of secondary data, namely $K = 2N, 1.5N, 2.5N$ are considered. From such figures, it is apparent that the proposed detectors and the tunable competitors (Kalson’s and KWA) have essentially the same $P_d$ of Kelly’s detector, while as known the AMF exhibits a performance loss, which becomes negligible only when the number of secondary data is large enough. Notice that the KWA for $\gamma = 0.4$ and the detector (15) for $\epsilon = 0.2$ start to experience some loss, especially for moderate SNR, hence smaller values of their respective parameters are not considered for a fair comparison.

Fig. 3 reports the results under mismatch, in particular considering a cosine squared of the angle between the nominal steering vector and the mismatched one (given by equation (B.26)) equal to $\cos^2 \theta = 0.46$ (obtained for $\delta = 0.4/N = 0.025$). It emerges that the proposed approach yields a family of robust receivers, whose performance in terms of robustness to mismatches depends on the design matrix $\Sigma$ and number of secondary data $K$. The behavior of the GLRT for $\Sigma = \mathbf{C}$ is very interesting,
because it is (equivalent to Kelly’s detector under matched conditions and) more robust than Kalson’s and KWA receivers; the parametric detector (15) with both settings of $\epsilon$ is more robust than the GLRT for $\Sigma = C$ and can be even more robust than AMF, in particular for $K = 32$, without the loss experienced by the latter under matched conditions. This “win-win” behavior is thus remarkably unique and not achievable with state-of-the-art receivers. Notice that the AMF can be considered a competitor only for $K = 40$ (where its performance loss under matched conditions compared to Kelly’s detector becomes very small), but Fig. 3(c) shows that the proposed detectors, in particular the GLRT for $\Sigma = C$ and the parametric detector (15) with both settings of $\epsilon$, are more robust than AMF; notice that this advantage comes at no increase in computational complexity.

The GLRT with $\Sigma = v_mv_m^H$ needs higher values of $K$ to achieve almost the same robustness of the other proposed receivers (Fig. 3(c)); otherwise, it progressively rejects the $H_1$ hypothesis: for $K = 32$ (Fig. 3(b)) only for larger SNR (not shown in the figure), for $K = 24$ (Fig. 3(a)) behaving as a selective receiver over the whole range of SNRs. As to Kalson’s detector, for $\beta = 0$ it is close to Kelly’s detector (in particular for $K = 24$).

For a reduced value of the mismatch, the differences between the proposed detectors and the competitors reduce. Fig. 4 shows the mismatched case with $\cos^2 \theta \approx 0.83$ (obtained for $\delta_f = 0.2/N = 0.0125$). We observe that the parametric detector (15) still emerges as the most robust among the detectors that have negligible loss under matched conditions (compared to Kelly’s...
We have proposed a novel family of robust radar receivers to detect the possible presence of a coherent return from a given cell under test mismatched with respect to the nominal steering vector. To this end, we have introduced a random component in addition to the deterministic useful signal; the random component has a given structure $\Sigma$, but its power $\nu$ is estimated from the data. This likely produces an estimate of $\nu$ close to zero under matched conditions, thus achieving no loss compared to Kelly’s detector (which is the benchmark for the considered problem).

We have solved the hypothesis testing problem resorting to the GLRT approach and assuming for $\Sigma$ either a known rank-one matrix or the unknown covariance matrix of the noise. We have also introduced a parametric detector that naturally modifies the statistic of the latter GLRT and shares with it the CFAR property. The performance analysis has shown that proposed detectors are a viable means to deal with mismatched signals; in fact, compared to competitors that have negligible loss under matched conditions, the proposed approach guarantees an increased robustness.

Outgrowths of this work are the investigation of further possible structures for the design matrix $\Sigma$, and possibly the derivation of the one-step GLRT for the general case (which in fact is missing in Table I). A different avenue for possible generalizations is robust detection of range-spread targets; to date, the GLRT for the proposed random perturbation approach has been derived only in the case $\Sigma = C$, which though has the same remarkable properties of the point-like detector derived here [30]. Finally, it would be interesting to ascertain whether receivers with the same detection power of Kelly’s detector (under matched conditions) but controllable robustness or selectivity under mismatched conditions can be obtained through design approaches different from GLRT, namely inspired to machine learning tools.

**APPENDIX A**

**DERIVATION OF EQUATION (9)**

First observe that, by the matrix inversion lemma, we have

$$u^\dagger \left( S + (1 + b)^{-1} (z - \alpha v)(z - \alpha v)^\dagger \right)^{-1} u = u^\dagger S^{-1} u - \frac{(1 + b)^{-1} \left| u^\dagger S^{-1} (z - \alpha v) \right|^2}{1 + (1 + b)^{-1} \left| (z - \alpha v) S^{-1} (z - \alpha v) \right|^2}.$$

Thus, $l(b, \alpha)$ can be re-written as

$$l(b, \alpha) = \frac{(K+1)^N\pi e}{\det^{K+1}(S) \left[ 1 + (\tilde{z} - \alpha \tilde{v})^\dagger (\tilde{z} - \alpha \tilde{v}) \right]^{K+1}} \times \left[ \frac{u^\dagger \tilde{u} - (1 + b)^{-1} \left| u^\dagger (\tilde{z} - \alpha \tilde{v}) \right|^2}{1 + (1 + b)^{-1} \left| (\tilde{z} - \alpha \tilde{v})^\dagger (\tilde{z} - \alpha \tilde{v}) \right|^2} \right]^{K+1},$$

where $\tilde{z} = S^{-1/2} z$, $\tilde{v} = S^{-1/2} v$, and $\tilde{u} = S^{-1/2} u$ are the whitened versions of $z, v,$ and $u$, respectively.
Notice also that
\[
\tilde{u}^\dagger \tilde{u} - (1 + b)^{-1} \frac{|\tilde{u}^\dagger (\tilde{\alpha} - \tilde{\nu})|^2}{1 + (1 + b)(\tilde{\alpha} - \tilde{\nu})(\tilde{\alpha} - \tilde{\nu})} = \frac{\tilde{u}^\dagger \tilde{u} - |\tilde{u}^\dagger (\tilde{\alpha} - \tilde{\nu})|^2}{1 + b(\tilde{\alpha} - \tilde{\nu})(\tilde{\alpha} - \tilde{\nu})} = \frac{\tilde{u}^\dagger \tilde{u} - |\tilde{u}^\dagger (\tilde{\alpha} - \tilde{\nu})|^2}{1 + b(\tilde{\alpha} - \tilde{\nu})(\tilde{\alpha} - \tilde{\nu})}
\]
\[
= (1 + b)\tilde{u}^\dagger \tilde{u} + \tilde{u}^\dagger \tilde{u} \|\tilde{\alpha} - \tilde{\nu}\|^2 - |\tilde{u}^\dagger (\tilde{\alpha} - \tilde{\nu})|^2 \times \frac{1 + (\tilde{\alpha} - \tilde{\nu})^\dagger (\tilde{\alpha} - \tilde{\nu})}{1 + b(\tilde{\alpha} - \tilde{\nu})(\tilde{\alpha} - \tilde{\nu})}
\]
It turns out that \(l(b, \alpha)\) can be re-cast as
\[
l(b, \alpha) = \frac{(K + 1)^N(1 + b)^{-1}}{\det K^1(S)} \left(1 + b + (\tilde{\alpha} - \tilde{\nu})^\dagger (\tilde{\alpha} - \tilde{\nu})\right)^{K + 1}
\]
\[
\times \left[1 + (1 + b)\|\tilde{u}\|^2 + \|\tilde{u}\|\|\tilde{\alpha} - \tilde{\nu}\|^2 - |\tilde{u}^\dagger (\tilde{\alpha} - \tilde{\nu})|^2\right]^{K + 1}
\]
and letting \(|\tilde{u}^\dagger (\tilde{\alpha} - \tilde{\nu})|^2 = \|\tilde{u}\|^2\|P_{\tilde{u}} (\tilde{\alpha} - \tilde{\nu})\|^2\) it can be re-written in the form of equation (9).

**Proof of Proposition 2**

First observe that \(l(b, \alpha)\) can be written as
\[
l(b, \alpha) = \frac{l_1(b, \|\tilde{\alpha} - \tilde{\nu}\|^2)}{\det K^1(S)} = \frac{l_2(b, \|P_{\tilde{u}} (\tilde{\alpha} - \tilde{\nu})\|^2)}{(K + 1)^{-N(K + 1)}}
\]
with
\[
l_1(b, y_1) = \frac{(1 + b)^{-1}}{1 + b + y_1^{K + 1}}, \quad l_2(b, y_2) = \frac{(1 + b + y_2)^{K + 1}}{1 + y_2}.
\]
In particular, \(l_1\) and \(l_2\) are strictly decreasing function of \(y_1\) and \(y_2\), respectively. Moreover, \(l_1\) attains its maximum at
\[
\alpha_1 = \arg \min_{\alpha} \|\tilde{\alpha} - \tilde{\nu}\| \tab (\tilde{\alpha} - \tilde{\nu}) = (\tilde{v}^\dagger \tilde{v})^{-1} \tilde{v} \tilde{z}
\]
and
\[
\|\tilde{\alpha} - \tilde{\nu}\|^2 = \|\tilde{\alpha} - \tilde{\nu}\|^2 = \|\tilde{\nu} - (\tilde{\alpha} - \tilde{\nu})\| \tab \tilde{v} \tilde{\nu}
\]
where \(\tilde{\nu} = \alpha_1 \tilde{\nu} + |\alpha_1 - 1|^2 \tilde{v} \tilde{\nu} \geq \tilde{v} \tilde{P}_{\tilde{v}}^\dagger \tilde{z} \quad (A.18)
\]
Thus, \(\forall y \geq \tilde{v} \tilde{P}_{\tilde{v}}^\dagger \tilde{z} = \tilde{v} \tilde{P}_{\tilde{v}}^\dagger \tilde{v} \tilde{\nu}^{-1} \tilde{v} \tilde{P}_{\tilde{v}}^\dagger \tilde{z}
\]
and is a circle centered at \(\alpha_1\) of proper radius. Similarly, \(l_2\) attains its maximum at
\[
\alpha_2 = \arg \min_{\alpha} \|P_{\tilde{u}} (\tilde{\alpha} - \tilde{\nu})\|^2
\]
and
\[
\|P_{\tilde{u}} (\tilde{\alpha} - \tilde{\nu})\|^2 = \|P_{\tilde{u}} (\tilde{\alpha} - \tilde{\nu})\|^2 = \|P_{\tilde{u}} (\tilde{\nu} - (\tilde{\alpha} - \tilde{\nu})\| \tab \tilde{v} \tilde{P}_{\tilde{v}}^\dagger \tilde{v}
\]
and
\[
\tilde{P}_{\tilde{u}}^\dagger (\tilde{\nu} - (\tilde{\alpha} - \tilde{\nu})\| \tab \tilde{v} \tilde{P}_{\tilde{v}}^\dagger \tilde{v}
\]
and is a circle centered at \(\alpha_2\) of proper radius.

Obviously, for \(\alpha_1 = \alpha_2\) the maximum of \(l(b, \alpha)\), given \(b\), is attained at \(\alpha_1\). Assuming instead \(\alpha_1 \neq \alpha_2\), we can prove that the values of \(\alpha\) maximizing \(l(b, \alpha)\) belong to the segment whose endpoints are \(\alpha_1\) and \(\alpha_2\), indicated hereafter as \(S_{\alpha_1, \alpha_2}\).
Thus, \( \alpha - z \leq \tilde{t} \), thus implying the derivative of \( f \).

It turns out that \( 1 + \|z\|^2 = 1 + z\bar{s}^{-1}z = (1 + \tilde{t}_K)/b \) and \( \|P_a \tilde{z}\|^2 = z\bar{s}^{-1}z - \frac{|z\bar{s}^{-1}v|^2}{v\bar{s}^{-1}v} = \frac{1}{b} - 1 \) and, hence, the decision statistic of the detector (15) can be re-written as

\[
\Lambda_c(z, S) = \begin{cases} 
\frac{1 + \tilde{t}_K}{b} \left( 1 - \frac{1}{\zeta_c} \right) \frac{1}{\zeta_c} & b < 1 - \frac{1}{\zeta_c} \\
1 + \tilde{t}_K, & \text{otherwise.}
\end{cases}
\]

Under the noise-only hypothesis the random variable (RV) \( \tilde{t}_K \), given by equation (B.21), is distributed according to a complex central F-distribution with 1 and \( K - N + 1 \) (complex) degrees of freedom and it is independent of \( b \), given by equation (B.22), which, in turn, obeys a complex central beta distribution with \( K - N + 2 \) and \( N - 1 \) (complex) degrees of freedom [3], [26], [27]. In symbols, we write \( \tilde{t}_K \sim CF_{1, K-N+1} \) and \( b \sim CB_{K-N+2, N-1} \). It is thus apparent that \( P_{fa} \) is independent of \( C \), and, hence, the detector possesses the CFAR property.

Moreover, we can compute \( P_{fa} \) as follows

\[
P_{fa} = \int_0^1 P[\Lambda_c(z, S) > \eta | b = x, H_0] f_b(x) \, dx
\]

where \( f_b(\cdot) \) denotes the PDF of the RV \( b \) [27], i.e.,

\[
f_b(x) = \frac{1}{B(K + 2 - N, N - 1)} x^{K-N+1} (1-x)^{N-2}
\]

with \( 0 \leq x \leq 1 \), while

\[
p(x) = P[\Lambda_c(z, S) > \eta | b = x, H_0] = \begin{cases} 
1 - F_{\tilde{t}_K}(y_c(x)), & 0 \leq x \leq 1 - \frac{1}{\zeta_c} \\
1 - F_{\tilde{t}_K}(\eta - 1), & 1 - \frac{1}{\zeta_c} \leq x \leq 1
\end{cases}
\]

with [27, formula (A.7)]

\[
1 - F_{\tilde{t}_K}(y_c) = \begin{cases} 
1, & y_c \geq 0 \\
, & y_c < 0
\end{cases}
\]

and \( y_c(\eta) = \eta x \left( \frac{\zeta_c - 1}{\zeta_c} \right) = \eta - 1 \). To compute the \( P_{fa} \), we observe that the derivative of \( y_c(\eta) \) is given by

\[
\frac{dy_c(x)}{dx} = \eta \left( \frac{1 - x}{x} \right)^{\frac{\zeta_c - 1}{\zeta_c}} \left( \frac{\zeta_c - 1}{\zeta_c} - \frac{1 - x}{x} \right)
\]

and it is positive for \( 0 < x < 1 - \frac{1}{\zeta_c} \). Moreover, \( \lim_{x \to 0^+} y_c(x) = -1 \) and \( y_c(\frac{1}{\zeta_c}) = -1 \). It turns out that \( P_{fa} = 1 \) for \( \eta \leq 1 \). If, instead, \( \eta > 1 \), denoting by \( \tilde{\tau}_a(\eta) \) that number in \((0,1-\frac{1}{\zeta_c})\) such that \( y_c(\tilde{\tau}_a) = 0 \), it follows that

\[
P_{fa} = \int_0^{\tilde{\tau}_a(\eta)} f_b(x) \, dx + \int_{\tilde{\tau}_a(\eta)}^{1-\frac{1}{\zeta_c}} \frac{1}{y_c(\eta)} f_b(x) \, dx + \frac{1}{\eta^{K-N+1}} \int_{1-\frac{1}{\zeta_c}}^{1} f_b(x) \, dx
\]

i.e., introducing \( c = \frac{1}{B(K+2-N,N-1)} = \frac{K_1}{K(N-2)} \),

\[
P_{fa} = c \int_0^{\tilde{\tau}_a(\eta)} x^{K-N+1} (1-x)^{N-2} \, dx
\]
\[ + \frac{a c}{\eta^{K+1-N}} \int_{\pi_1(n)}^{1} x^{K+1-N} e^{-x} (1-x)^{N-2-K+1-N} \frac{dx}{\zeta} \]

\[ + \frac{c}{\eta^{K+1-N}} \int_{1}^{1} x^{K-N+1} (1-x)^{N-2} \frac{dx}{\zeta} \]

where \( a = \frac{(\zeta-1)^{K+1-N}}{\zeta^{K+1-N}} \). The \( P_{fa} \) of (15) is thus given by

\[ P_{fa} = cB_{\pi_1(n)}(K-N+2,N-1) + \frac{a c}{\eta^{K+1-N}} \]

\[ \times \left( B_1 - \frac{K+1-N}{\zeta} + 1, N-1 - \frac{K+1-N}{\zeta} \right) \]

\[ - B_{\pi_1(n)} \left( \frac{K+1-N}{\zeta} + 1, N-1 - \frac{K+1-N}{\zeta} \right) \]

\[ + \frac{c}{\eta^{K+1-N}} \left[ B(K-N+2,N-1) \right] \]

\[ (B.24) \]

where \( B(\cdot, \cdot) \) is the Eulerian beta function while \( B(\nu, \mu) = \int_{0}^{\infty} \frac{t^{\nu-1} (1-t)^{\mu-1}}{\Gamma(\nu) \Gamma(\mu)} dt \) is the incomplete beta function [28].

As a special case, i.e., \( \epsilon = 0 \) and, hence, \( \zeta = \zeta \), we obtain the \( P_{fa} \) of the GLRT (12).

On the other hand, if we suppose that under the \( H_1 \) hypothesis the actual useful signal is deterministic, but with a steering vector \( p \) different from the nominal one, \( v \), i.e.,

\[ r = \alpha p + n, \quad n \sim CN(0, C) \]

\[ \hat{t}_K, \text{ given } b, \text{ is ruled by a complex noncentral F-distribution with } K-N+1 \text{ (complex) degrees of freedom and noncentrality parameter } \delta, \text{ with } \]

\[ \delta^2 = SNR \cdot b \sin^2 \theta, \]

where

\[ SNR = |\alpha|^2 | p | C^{-1} | p | \]

and

\[ \cos^2 \theta = \frac{| p | C^{-1} | p |^2}{( p | C^{-1} | p |)^2}. \]

In addition, the RV \( b \) obeys the complex noncentral beta distribution with \( K-N+2 \) and \( N-1 \) (complex) degrees of freedom and noncentrality parameter \( \delta_b \), with

\[ \delta_b^2 = SNR \cdot \sin^2 \theta. \]

In symbols, we write \( \hat{t}_K \sim CF_{1,K-N+1}(\delta) \) and \( b \sim C \beta_{K-N+2,N-1}(\delta_b) \).

In the special case of perfect match between \( p \) and \( v \), then \( \delta^2 = |\alpha|^2 v \cdot C^{-1} v \cdot b \) and \( \delta^2 = 0 \), i.e., \( b \sim C \beta_{K-N+2,N-1} \) and \( \hat{t}_K \sim CF_{1,K-N+1}(\delta) \) given \( b \).

The above characterization highlights that performances of detectors (12) and (15) can be expressed in terms of \( P_d \) vs SNR (equation (B.25)) given \( \cos^2 \theta \) (equation (B.26)) and \( P_{fa} \). In principle, such characterization could be exploited to compute \( P_d \), parallelizing the computation of \( P_{fa} \); however, analytical forms for \( P_d \) are less useful since Monte Carlo simulation is not very time-consuming. As a matter of fact, we use Monte Carlo simulation to compute \( P_d \) vs SNR, expressed by equation (B.25), given \( \cos^2 \theta \) and \( P_{fa} \).

**APPENDIX C**

**THE TWO-STEP GLRT-BASED DETECTOR WITH**

\( \Sigma = vv^\dagger \) IS THE AMF

We focus on the binary hypotheses testing problem

\[ \left\{ \begin{array}{l}
H_0 : \quad z \sim CN(0, C) \\
H_1 : \quad z \sim CN(\alpha v, \nu vv^\dagger + C)
\end{array} \right. \]

where \( \nu \geq 0 \) and \( \alpha \in C \) are unknown quantities while the vectors \( u, v \in C^{N \times 1} \) and the Hermitian positive definite matrix \( C \) are known. The corresponding GLRT is given by

\[ \Lambda(z, C) = \max_{\nu \geq 0} \max_{\alpha \in C} f_1(z|\nu, \alpha) \frac{H_1}{H_0} \geq \eta \]  

where

\[ f_1(z|\nu, \alpha) = \frac{1}{\pi^N (1 + \nu vv^\dagger) \det(C)} \]

is the PDF of \( z \) under \( H_1 \) with

\[ f_0(z) = \frac{1}{\pi^N \det(C)} e^{-z^\dagger C^{-1} z}. \]

To maximize \( l(\nu, \alpha) \) over \( \alpha \) we rewrite its exponent as

\[ -(z - \alpha v)^\dagger C^{-1} (z - \alpha v) + \nu \frac{|\alpha|^2 | u | C^{-1} (z - \alpha v)|^2}{1 + \nu vv^\dagger C^{-1} u} \]

\[ = - \|z - \alpha v\|^2 + \nu \|\tilde{u} (\bar{z} - \alpha \bar{v})\|^2 \]

where \( \bar{z} = C^{-1/2} z, \tilde{v} = C^{-1/2} v, \) and \( \bar{u} = C^{-1/2} u \). Using

\[ \|z - \alpha v\|^2 = \|P_{\tilde{u}} (\bar{z} - \alpha \bar{v})\|^2 + \|P_{\tilde{u}} (\bar{z} - \alpha \bar{v})\|^2 \]

and

\[ \|\tilde{u} (\bar{z} - \alpha \bar{v})\|^2 \]

we obtain

\[ l(\nu, \alpha) = e \frac{- \|z - \alpha v\|^2 + \nu \|\tilde{u}\|^2 \|P_{\tilde{u}} (\bar{z} - \alpha \bar{v})\|^2}{1 + \nu \|\tilde{u}\|^2} \]

Now letting \( u = v \) and, hence, \( \tilde{u} = \tilde{v} \) we have an expression that can be easily maximized with respect to \( \alpha \) obtaining

\[ \max_{\alpha \in C} f_1(z|\nu, \alpha) = \frac{1}{\pi^N (1 + \nu vv^\dagger) \det(C)} e^{-\|P_{\tilde{u}} \bar{z}\|^2} \]

and hence

\[ \max_{\nu \geq 0} \max_{\alpha \in C} f_1(z|\nu, \alpha) = \frac{1}{\pi^N \det(C)} e^{-\|P_{\tilde{u}} \bar{z}\|^2} \]

It follows that

\[ A(z, C) = e^{\|P_{\tilde{u}} \bar{z}\|^2} = e^{-\nu vv^\dagger} \frac{H_1}{H_0} \geq \eta. \]

\[ (C.28) \]
The corresponding adaptive detector, obtained by replacing $C$ with the sample covariance matrix based on the secondary data, is obviously equivalent to the AMF.

REFERENCES


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